MoM Analysis of an Axisymmetric Chiral Radome

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Abstract – An axisymmetric chiral radome has been analyzed numerically by using the method of moments. The chiral body is illuminated by a plane wave and the surface equivalence principle is used to replace the body by equivalent electric and magnetic surface currents. The effect of adding chirality to a dielectric radome of revolution is investigated throughout numerical results obtained for bodies of different shapes and material parameters. Chiral materials can be used to design anti-reflective structures to control scattering cross section patterns of bodies. A computer program is developed for the chiral radome of revolution and examples of numerical calculations are given for a chiral spherical radome, a chiral cylindrical radome, and a chiral Von Karman radome. Numerical results for the chiral spherical radome are in excellent agreement with the exact ones obtained by the eigenfunction solution. Moreover, the numerical results of the chiral Von Karman radome are in excellent agreement with the published results.

Index Terms - Axisymmetric radome, chiral radome, method of moments, and surface equivalence theorem.

I. INTRODUCTION

An axisymmetric chiral radome has been analyzed numerically by using Method of Moments (MoM) with the surface equivalence principle. Scattering and radiation from radomes and antenna systems with radomes have been studied in the last couple decades by using different methods, such as the ray tracing technique [1-3], the plane wave spectrum-surface integral technique [4], the MoM [5-8], the physical optics (PO) method and dielectric physical optics (DPO) technique [9], the finite element method (FEM) [10], the method of regularization (MoR) [11], the hybrid PO-MoM technique [12], the transmission-line modeling method [13], the adaptive integral method [14], the dyadic Green's function (DGF) technique [15-16], the fast Fourier transform (FFT) [17] and precorrected fast Fourier transform methods (P-FFT) [18], body of revolution (BOR) formulations of MoM [19-26], and the finite difference time domain (FDTD) method [27].

Arbitrary dielectric bodies, 3-D arbitrary lossy dielectric bodies, arbitrary conducting bodies, dielectric bodies of revolution, conducting bodies of revolution with and without apertures, dielectric radomes of revolution, chiral and/or metal coated dielectric bodies, a 2-D chiral radome of arbitrary shape, and dielectric radomes with antenna systems were investigated in [1–35] by using the methods mentioned above.

We have not found any work that uses MoM to analyze an axisymmetric 3-D chiral radome and calculates the internal fields and scattering from it. Here, we consider different shapes of axisymmetric chiral radomes to find out the effects of sizes and shapes on the internal fields and scattered fields outside by using MoM with BOR formulations. This work is a continuation of our previous work [36].

II. ANALYSIS

A plane wave is incident on the homogeneous chiral shell of permittivity ε_2 , permeability μ_2 , and chirality ζ_2 , shown in Fig. 1, where the η 's are intrinsic impedances. A homogeneous region characterized by the medium parameters $\varepsilon_1 = \varepsilon_0$ and $\mu_1 = \mu_0$ surrounds the shell. ($\mathbf{E}^i, \mathbf{H}^i$) represents the incident field produced by external sources in the absence of the shell. S₁ and S₂ represent the outer and inner surfaces of the shell, respectively. The field (\mathbf{E}, \mathbf{H}) in the region bounded by S₂ and the scattered field external to S₁ are of interest in this paper.



Fig. 1. A chiral radome illuminated by a plane wave.

A. Surface equivalence of the problem

Using the equivalence principle, the problem of Fig. 1 can be reduced to three simpler and equivalent problems shown in Figs. 2, 3, and 4. In the external equivalence, electric surface current J_1 and magnetic surface current M_1 have been placed on S_1 . These surface currents are radiating in an unbounded medium of (ε_1 , μ_1) with the same incident field of Fig. 1. The total field at any point in the external region bounded by S_1 in Fig. 2 is the same as the total field at the same point of Fig. 1, while the total field at any point in the internal region bounded by S_1 in Fig. 2 is zero.

$$\mathbf{E}_{1 \tan}(\mathbf{J}_1, \mathbf{M}_1) = -\mathbf{E}_{\tan}^{\prime} \text{ on } \mathbf{S}_1^{-}$$
(1)

$$\mathbf{H}_{1 \tan} \left(\mathbf{J}_1, \mathbf{M}_1 \right) = -\mathbf{H}_{\tan}^{\iota} \text{ on } \mathbf{S}_1^{-}, \qquad (2)$$

$$\mathbf{J}_1 = \mathbf{n}_1 \times \mathbf{H}_{out}^+ \,, \tag{3}$$

$$\mathbf{M}_1 = \mathbf{E}_{out}^+ \times \mathbf{n}_1, \qquad (4)$$

where the superscript "–" on S_1 indicates the side of S_1 opposite the region into which n_1 points. $E_1(J_1, M_1)$ and $H_1(J_1, M_1)$, respectively, denote the electric and magnetic fields produced by the surface currents \mathbf{J}_1 and \mathbf{M}_1 when they radiate in the unbounded medium of (ε_l, μ_l) . \mathbf{n}_1 denotes the unit outward vector on S_1 and $(\mathbf{E}_{out}^+, \mathbf{H}_{out}^+)$ are the total fields just outside S_1 in Fig. 1.



Fig. 2. External equivalence of the problem shown in Fig. 1.

In the internal equivalence for the region bounded by S_1 and S_2 , surface currents $-J_1$, $-J_2$, $-M_1$, and $-M_2$ are placed on S_1 and S_2 where they radiate in the unbounded medium of (ε_2 , μ_2 , ξ_2). They produce the correct total fields (**E**, **H**) at any point in the region bounded by S_1 and S_2 and produce zero fields at any point outside the region bounded by S_1 and S_2 .

$$\mathbf{E}_{2 \tan}\left(\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2}\right) = 0 \text{ on } \mathbf{S}_{1}^{\top}$$
 (5)

$$\mathbf{E}_{2 \tan} \left(\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2} \right) = 0 \text{ on } \mathbf{S}_{2}^{+}, \qquad (6)$$

$$\mathbf{H}_{2 \tan}(\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2}) = 0 \text{ on } \mathbf{S}_{1}^{+}, \qquad (7)$$

$$\mathbf{H}_{2 \tan} \left(\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2} \right) = 0 \text{ on } \mathbf{S}_{2}^{+}, \quad (8)$$

$$\mathbf{J}_2 = \mathbf{n}_2 \times \mathbf{H}_{in}^-, \qquad (9)$$

$$\mathbf{M}_2 = \mathbf{E}_{in} \times \mathbf{n}_2, \tag{10}$$

where S_1^+ and S_2^+ denote the sides of the surfaces S_1 and S_2 facing the regions into which the unit vectors \mathbf{n}_1 and \mathbf{n}_2 point, and $\mathbf{E}_2(\mathbf{J}_1, \mathbf{M}_1, \mathbf{J}_2, \mathbf{M}_2)$ and $\mathbf{H}_2(\mathbf{J}_1, \mathbf{M}_1, \mathbf{J}_2, \mathbf{M}_2)$, respectively, denote the electric and magnetic fields produced by the equivalent surface currents when they radiate in the unbounded medium of (ε_2 , μ_2 , ξ_2). \mathbf{n}_1 and \mathbf{n}_2 denote the unit vectors on S_1 and S_2 , respectively and ($\mathbf{E}_{in}^-, \mathbf{H}_{in}^-$) are the total fields on S_2^- as shown in Fig. 1.

In the internal equivalence for the region bounded by S₂, surface currents J_2 , and M_2 are placed on S₂ where they radiate in the unbounded medium of (ε_1 , μ_1). They produce the correct total fields (**E**, **H**) at any point in the region bounded by S_2 and produce zero fields at any point outside the region bounded by S_2 ,

$$\mathbf{E}_{1 \tan} (\mathbf{J}_2, \mathbf{M}_2) = 0 \text{ on } \mathbf{S}_2^-$$
 (11)

$$\mathbf{H}_{1 \tan}(\mathbf{J}_2, \mathbf{M}_2) = 0 \text{ on } \mathbf{S}_2^{-}.$$
 (12)



Fig. 3. Internal equivalence for the region bounded by S_1 and S_2 of the problem shown in Fig. 1.



Fig. 4. Internal equivalence for the region bounded by S_2 of the problem shown in Fig. 1.

B. Formulation of the integral equations

Equations (1), (2), (5) - (8), (11), and (12) represent eight coupled integral equations for the four unknown surface currents J_1 , J_2 , M_1 , and M_2 . The combined field formulation reduces these eight equations to four by adding equations (1) to (5), (6) to (11), (2) to (7), and (8) to (12). These four coupled integral equations are solved numerically by using the method of moments.

$$-\mathbf{E}_{1\tan}(\mathbf{J}_1,\mathbf{M}_1)\Big|_{S_1}^{S_1} - \mathbf{E}_{2\tan}(\mathbf{J}_1,\mathbf{M}_1,\mathbf{J}_2,\mathbf{M}_2)\Big|_{S_1}^{S_1} = \mathbf{E}_{\tan}^i \quad (13)$$

$$-\mathbf{E}_{1 \tan} (\mathbf{J}_{2}, \mathbf{M}_{2}) \Big|_{52}^{52} - \mathbf{E}_{2 \tan} (\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2}) \Big|_{52}^{52} = 0, (14)$$

$$-\mathbf{H}_{1\tan}(\mathbf{J}_1,\mathbf{M}_1)|^{\sigma_1} - \mathbf{H}_{2\tan}(\mathbf{J}_1,\mathbf{M}_1,\mathbf{J}_2,\mathbf{M}_2)|^{\sigma_1} = \mathbf{H}_{\tan}^{\iota}, (15)$$

$$-\mathbf{H}_{1 \tan}(\mathbf{J}_{2}, \mathbf{M}_{2})|^{2} - \mathbf{H}_{2 \tan}(\mathbf{J}_{1}, \mathbf{M}_{1}, \mathbf{J}_{2}, \mathbf{M}_{2})|^{2} = 0.$$
(16)

The electric and magnetic fields produced by \mathbf{J} and \mathbf{M} in an unbounded chiral medium are given by [37, (1.2.4) and (1.2.5)].

C. Expansion functions and testing

Let the electric and magnetic surface currents J_1 , J_2 , M_1 , and M_2 be expanded as,

$$\mathbf{J}_{1} = \sum_{n=-\infty}^{\infty} \sum_{j=1}^{N_{1}} \left(I_{1nj}^{t} \mathbf{J}_{1nj}^{t} + I_{1nj}^{\phi} \mathbf{J}_{1nj}^{\phi} \right)$$
(17)

$$\mathbf{J}_{2} = \sum_{n=-\infty}^{\infty} \sum_{j=1}^{N_{2}} \left(I_{2nj}^{t} \mathbf{J}_{2nj}^{t} + I_{2nj}^{\phi} \mathbf{J}_{2nj}^{\phi} \right), \qquad (18)$$

$$\mathbf{M}_{1} = \eta_{1} \sum_{n=-\infty}^{\infty} \sum_{j=1}^{N_{1}} \left(V_{1nj}^{t} \mathbf{J}_{1nj}^{t} + V_{1nj}^{\phi} \mathbf{J}_{1nj}^{\phi} \right), \quad (19)$$

$$\mathbf{M}_{2} = \eta_{1} \sum_{n=-\infty}^{\infty} \sum_{j=1}^{N_{2}} \left(V_{2nj}^{t} \mathbf{J}_{2nj}^{t} + V_{2nj}^{\phi} \mathbf{J}_{2nj}^{\phi} \right).$$
(20)

 I_{1nj}^{t} , I_{1nj}^{ϕ} , I_{2nj}^{t} , I_{2nj}^{ϕ} , V_{1nj}^{t} , V_{1nj}^{ϕ} , V_{2nj}^{t} , and V_{2nj}^{ϕ} are coefficients to be determined. \mathbf{J}_{1nj}^{t} , \mathbf{J}_{1nj}^{ϕ} , \mathbf{J}_{2nj}^{t} , and \mathbf{J}_{2nj}^{ϕ} are given below,

$$\mathbf{J}_{1nj}^{t} = \mathbf{u}_{t} f_{1j}(t') e^{jn\phi'}$$
(21)

$$\mathbf{J}_{1nj}^{\phi} = \mathbf{u}_{\phi} f_{1j}(t') e^{jn\phi'}, \qquad (22)$$

$$\mathbf{J}_{2nj}^{t} = \mathbf{u}_{t} f_{2j}(t') e^{jn\phi'}, \qquad (23)$$

$$\mathbf{J}_{2nj}^{\phi} = \mathbf{u}_{\phi} f_{2j} \left(t' \right) e^{jn\phi'}, \qquad (24)$$

$$f_{1j}(t') = \frac{1}{\rho'} T(t' - \overline{t}_{1,2j+1}), \qquad (25)$$

$$f_{2j}(t') = \frac{1}{\rho'} T(t' - \overline{t}_{2,2j+1}), \qquad (26)$$

$$f_{1i}(t) = \frac{1}{\rho} T(t - \overline{t}_{1,2i+1}) , \qquad (27)$$

$$f_{2i}(t) = \frac{1}{\rho} T(t - \overline{t}_{2,2i+1}), \qquad (28)$$

$$T(t - \overline{t}_{(1,2)2i+1}) = \sum_{p=1}^{4} T_{(1,2)p+4i-4} \ \delta \ (t - t_{(1,2)p+2i-2}), \ (29)$$

where $\delta(t)$ is the unit impulse function. The right-hand side of equation (29) is the four impulse approximation to a triangle function shown in Fig. 5. Also, *t* is the arc length along the generating curve of either S₁ or S₂, ϕ is the angle of the line from the origin to a specified point in the *xy*-plane with respect to the *x*-axis, ρ is the distance from the *z*-axis, and u_t and u_{ϕ} are the unit vectors in the *t*- and ϕ - directions, respectively. And dropping the subscripts 1 and 2 in equations (21) - (29),

$$T_{4i-3} = \frac{d_{2i-1}^2}{2(d_{2i-1} + d_{2i})}$$
(30)

$$T_{4i-2} = \frac{(d_{2i-1} + \frac{1}{2}d_{2i})d_{2i}}{d_{2i-1} + d_{2i}},$$
 (31)

$$T_{4i-1} = \frac{(d_{2i+2} + \frac{1}{2}d_{2i+1})d_{2i+1}}{d_{2i+1} + d_{2i+2}},$$
 (32)

$$T_{4i} = \frac{d_{2i+2}^2}{2(d_{2i+1} + d_{2i+2})} \,. \tag{33}$$

An odd number greater than or equal to 5 of consecutive points $\overline{t_i}$ at $(\overline{\rho_i}, \overline{z_i})$, i = 1, 2, ..., P on the generating curves of the surfaces of revolution are defined. The generating curves are approximated by drawing straight lines between the points $(\overline{\rho_i}, \overline{z_i})$, i = 1, 2, ..., P and we define

$$d_{i} = \sqrt{\left(\overline{\rho}_{i+1} - \overline{\rho}_{i}\right)^{2} + \left(\overline{z}_{i+1} - \overline{z}_{i}\right)^{2}}.$$
 (34)



Fig. 5. Triangle function $T(t - \overline{t}_{2i+1})$ and four impulse approximation.

The quantities in equations (30) - (34) need to be specified to either the generating curve of S_1 or the generating curve of S_2 . If both **h** and **g** are vector functions on S_1 or if both **h** and **g** are vector functions on S_2 , then the symmetric product of **g** with **h** is $\langle \mathbf{h}, \mathbf{g} \rangle$ defined as

$$\langle \mathbf{h}, \mathbf{g} \rangle = \int_{S} \mathbf{h} \cdot \mathbf{g} \, dS,$$
 (35)

where S is S_1 if both **g** and **h** are on S_1 and S is S_2 if both **g** and **h** are on S_2 .

By taking the symmetric products of equations (13) and (15) with \mathbf{J}_{-1ni}^{t} and \mathbf{J}_{-1ni}^{ϕ} , $\mathbf{i} = 1, 2, ..., N_{1}$, and taking the symmetric products of equations (14) and (16) with \mathbf{J}_{-2ni}^{t} and \mathbf{J}_{-2ni}^{ϕ} , $\mathbf{i} = 1, 2, ..., N_{2}$, we obtain a matrix equation for the unknown coefficients. An incident plane wave whose propagation vector is in the *xz*-plane is considered.

D. Scattered field far from the scatterer and scattering cross section

The scattered field far from the scatterer is obtained by using the reciprocity theorem [38, Section 3-8]. For $p = \theta$ or ϕ , the *p*-component of the scattered field at the location \mathbf{r}^{rec} of the receiver is found after some calculations to be [37, 4.6.3]

$$E_{pq}^{scat} = -\frac{j\eta_{1}e^{-jk_{1}r^{rec}}}{4\pi r^{rec}} \sum_{n=-N}^{N} \left(\tilde{R}_{n}^{p}T_{n}^{q}\right) e^{jn\phi^{rec}}, \quad p,q=\theta, \ \phi, \ (36)$$

where r^{rec} is the distance from the origin in the vicinity of the scatterer to \mathbf{r}^{rec} . The extra subscript q in E_{pq}^{scat} is θ for the θ -polarized incident electric field and ϕ for the ϕ -polarized incident electric field. In equation (36), the i^{th} element of T_n^q is the coefficient that multiplies the i^{th} of the $e^{\text{jn}\phi}$ -dependent expansion functions for the equivalent electric and magnetic currents in the expressions for the equivalent electric and magnetic currents that radiate the far field. The contribution of the i^{th} element of \tilde{R}_n^p to the far field is the right-hand side of equation (36) with the summation with respect to n replaced by the product of the three quantities which are the i^{th} element of \tilde{R}_n^p , the i^{th}

The scattering cross section σ_{pq} is the area by which the power per unit area of the incident plane wave whose electric field is *q*-polarized must be multiplied to obtain, by isotropic radiation, the power per unit area of the *p*-component E_{pq}^{scat} of the scattered electric field. Because the isotropic radiator of power *P* produces the power per unit area $P/(4\pi (r^{rec})^2)$ at the distance r^{rec} where a receiver is located, this isotropic radiator will produce the power per unit area $|E_{pq}^{scat}|^2/\eta_1$ of the *p*component of the scattered electric field at the distance r^{rec} if

$$\frac{P}{4\pi \left(r^{rec}\right)^2} = \frac{\left|E_{pq}^{scat}\right|^2}{\eta_1},$$
(37)

where η_1 is the intrinsic impedance of the medium. Using the definition of σ_{pq} to set *P* equal to the product of σ_{pq} with the incident power per

unit area $|\mathbf{E}^{q}|^{2}/\eta_{1}$ of the q-polarized incident electric field \mathbf{E}^{q} , one obtains, after some calculations [37, 4.7.5],

$$\frac{\sigma_{pq}}{\lambda_1^2} = \frac{\left|\sum_{n=-N}^{N} \left(\tilde{R}_n^p T_n^q\right) e^{jn\phi^{rec}}\right|^2}{16\pi^3} \text{ where } \lambda_1 = \frac{2\pi}{k_1}.(38)$$

E. Electromagnetic field inside the radome

The electromagnetic field inside the region bounded by S_2 is calculated by using the combination of surface currents J_2 and M_2 , radiating in all space filled with the homogeneous medium (ε_1 , μ_1). The method of moment solutions for J_2 and M_2 are given by equations (18) and (20), respectively.

The electromagnetic field inside the region bounded by S_2 is calculated by,

$$\mathbf{E}_{i} = \sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(I_{2nj}^{t} \mathbf{E}_{i} \left(\mathbf{J}_{2nj}^{t}, \mathbf{0} \right) + I_{2nj}^{\phi} \mathbf{E}_{i} \left(\mathbf{J}_{2nj}^{\phi}, \mathbf{0} \right) + \eta_{1} \left(V_{2nj}^{t} \mathbf{E}_{i} \left(\mathbf{0}, \mathbf{J}_{2nj}^{t} \right) + V_{2nj}^{\phi} \mathbf{E}_{i} \left(\mathbf{0}, \mathbf{J}_{2nj}^{\phi} \right) \right) \right)$$
(39)

$$\mathbf{H}_{i} = \sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(I_{2nj}^{t} \mathbf{H}_{i} (\mathbf{J}_{2nj}^{t}, \mathbf{0}) + I_{2nj}^{\phi} \mathbf{H}_{i} (\mathbf{J}_{2nj}^{\phi}, \mathbf{0}) + \eta_{1} \left(V_{2nj}^{t} \mathbf{H}_{i} (\mathbf{0}, \mathbf{J}_{2nj}^{t}) + V_{2nj}^{\phi} \mathbf{H}_{i} (\mathbf{0}, \mathbf{J}_{2nj}^{\phi}) \right) \right), \quad (40)$$

where the subscript *i* in \mathbf{E}_i and \mathbf{H}_i indicates the radiation in all space filled with the medium that is bounded by \mathbf{S}_2 . The first argument of each of \mathbf{E}_i and \mathbf{H}_i is treated as an electric current and the second argument is treated as a magnetic current.

Using
$$\mathbf{E}_i(\mathbf{0}, \mathbf{J}) = -\mathbf{H}_i(\mathbf{J}, \mathbf{0})$$
 and $\mathbf{H}_i(\mathbf{0}, \mathbf{J}) = \frac{1}{\eta_1^2} \mathbf{E}_i(\mathbf{J}, \mathbf{0})$

to reduce all nonzero magnetic current arguments in equations (39) and (40) to zeros and then suppressing all the zero magnetic current arguments, one obtains

$$\mathbf{E}_{i} = \sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(I_{2nj}^{t} \mathbf{E}_{i} \left(\mathbf{J}_{2nj}^{t} \right) + I_{2nj}^{\phi} \mathbf{E}_{i} \left(\mathbf{J}_{2nj}^{\phi} \right) -\eta_{1} \left(V_{2nj}^{t} \mathbf{H}_{i} \left(\mathbf{J}_{2nj}^{t} \right) + V_{2nj}^{\phi} \mathbf{H}_{i} \left(\mathbf{J}_{2nj}^{\phi} \right) \right) \right), \qquad (41)$$

$$\mathbf{H}_{i} = \sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(I_{2nj}^{t} \mathbf{H}_{i} (\mathbf{J}_{2nj}^{t}) + I_{2nj}^{\phi} \mathbf{H}_{i} (\mathbf{J}_{2nj}^{\phi}) + \frac{1}{\eta_{1}} \left(V_{2nj}^{t} \mathbf{E}_{i} (\mathbf{J}_{2nj}^{t}) + V_{2nj}^{\phi} \mathbf{E}_{i} (\mathbf{J}_{2nj}^{\phi}) \right) \right).$$
(42)

Define

$$\mathbf{Z}_{nj}^{q} = -\frac{1}{\eta_{1}} \mathbf{E}_{i} (\mathbf{J}_{2nj}^{q}), \quad q = t, \ \phi, \qquad (43)$$

$$\mathbf{Y}_{nj}^{q} = -\mathbf{H}_{i}(\mathbf{J}_{2nj}^{q}), \quad q = t, \ \phi,$$
(44)

and use equations (43) and (44) in (41) and (42) to obtain

$$\mathbf{E}_{i} = -\sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(\eta_{1} \left(I_{2nj}^{t} \mathbf{Z}_{nj}^{t} + I_{2nj}^{\phi} \mathbf{Z}_{nj}^{\phi} \right) -\eta_{1} \left(V_{2nj}^{t} \mathbf{Y}_{nj}^{t} + V_{2nj}^{\phi} \mathbf{Y}_{nj}^{\phi} \right) \right),$$
(45)
$$\mathbf{H}_{i} = -\sum_{n=-N}^{N} \sum_{j=1}^{N_{2}} \left(\left(I_{2nj}^{t} \mathbf{Y}_{nj}^{t} + I_{2nj}^{\phi} \mathbf{Y}_{nj}^{\phi} \right) + \left(V_{2nj}^{t} \mathbf{Z}_{nj}^{t} + V_{2nj}^{\phi} \mathbf{Z}_{nj}^{\phi} \right) \right).$$
(46)

III. NUMERICAL RESULTS

The bodies analyzed in this paper, shown in Fig. 6, are illuminated by the θ -polarized plane waves. For $\theta^{inc} = 180^{\circ}$, the θ -polarized plane wave travels in the z-direction and its electric field is in the *-x*-direction. For $\theta^{inc} = 0^{\circ}$, the θ -polarized plane wave travels in the *-z*-direction and its electric field is in the *x*-direction.



Fig. 6. Structures used to verify the method.

Numerical solution results for inside fields on the z-axis and scattering from a spherical chiral radome are presented in Figs. 7 and 8 for relative permittivity $\varepsilon_r = 2$ and 8 and relative chirality $\xi_r = 0.4$. The internal field plots are normalized to the incident field. The computed results are compared with the exact solution results for the spherical chiral radome and they match perfectly with each other. For $\varepsilon_r = 2$ problem, 89 triangles on S₁ and 79 triangles on S₂ are used to solve for 672 unknowns, which took 6.6 minutes on a Core2Duo 2.1GHz computer. For $\varepsilon_r = 8$ problem, 179 triangles on S₁ and 159 triangles on S₂ are used to solve for 1352 unknowns, which took 21 minutes on the same computer.



Fig. 7. Fields on the *z*-axis of a spherical radome for R1 = $1\lambda_1$, R2 = $0.9\lambda_1$, $\mu_r = 1$, and $\theta^{inc} = 180^\circ$.



Fig. 8. $\sigma_{\theta\theta}$ of a spherical radome for R1 = $1\lambda_1$, R2 = $0.9\lambda_1$, $\mu_r = 1$, and $\theta^{inc} = 180^{\circ}$.

Figures 9 and 10 show numerical results for a cylindrical radome of R1 = $1.05\lambda_1$, R2 = $1\lambda_1$, H1 = $10.1\lambda_1$, H2 = $10\lambda_1$, μ_r = 1, and $\theta^{inc} = 0^\circ$. For $\varepsilon_r = 1.55$ problem, 277 triangles on S₁ and 273 triangles on S₂ are used to solve for 2212 unknowns. For $\varepsilon_r = 3$ problem, 306 triangles on S₁ and 299 triangles on S₂ are used to solve for 2432 unknowns.



Fig. 9. Fields on the *z*-axis of a cylindrical radome for $\theta^{inc} = 0^{\circ}$.



Fig. 10. $\sigma_{\theta\theta}$ of a cylindrical radome for $\theta^{inc} = 0^{\circ}$.

Results for small Von Karman radomes of different sizes and parameters were computed and validated against [33] - [35]. In Figs. 11 to 13, we show computed numerical results for a Von Karman radome of $\varepsilon_r = 4$, $L1 = 2\lambda_1$, $L2 = 1.8\lambda_1$, $D1 = 1\lambda_1$, $D2 = 0.9\lambda_1$, $\xi_r = 0$, $\mu_r = 1$ where 127 triangles are used on **S**₁ and 115 triangles are used on **S**₂ to solve for 980 unknowns.



Fig. 11. Fields on the *z*-axis of a Von Karman radome for $\theta^{inc} = 0^{\circ}$, 50°, and 90°.



Fig. 12. Fields on the *z*-axis of a Von Karman radome for $\theta^{inc} = 0^{\circ}$, 50°, and 90°.



Fig. 13. $\sigma_{\theta\theta}$ of a Von Karman radome for $\theta^{inc} = 0^{\circ}$, 50°, and 90°.

More results for all these different radome structures and comparisons with previously published results are available in [37].

IV. CONCLUSION

this paper, MoM analysis of an In axisymmetric chiral radome using MoM with the surface equivalence principle is presented. The body is replaced by equivalent electric and magnetic surface currents, which produce the correct fields inside and out. The application of the boundary conditions on the tangential components of the total electric and the total magnetic fields results in a set of eight equations, which then reduces to four coupled equations that needs to be solved. Triangular expansion functions are used for both *t*-directed and ϕ -directed currents. The unknown coefficients of these expansion functions are obtained using the method of moments.

The surface currents, inside fields and the scattering cross section are computed. The results are generated by a computer code, which produces excellent agreement with the exact solution for the spherical radome and agreement with available published results for other radomes. Increasing the number of segments increases the accuracy of the solution. The radar cross section (RCS) of the radome is useful because it tells how visible the radome is from the outside. The field inside the radome due to a plane wave incident on the radome is more useful because it tells how the radome distorts radiation that comes from outside the radome. If a receiver is placed inside the radome, the field inside the radome tells how the radome affects what is received from outside. Therefore, calculations of the RCS of the radome and fields inside the radome are justified. The field radiated outside the radome by a transmitter inside the radome is also of interest. This field was not computed because of its excitation, which is the incident field of the transmitter inside the radome, is more complicated than the incident plane wave excitation used to compute the field inside the radome.

The presence of the radome affects what is transmitted and received by a transceiver placed inside the radome. Adding chirality to the radome material affects the inside fields and the scattered fields significantly for the cases studied in this paper. Although the change from no chirality to a small chirality causes cross polarized field components to evolve, thereby shifting the directions of the scattered field outside the radome and the field inside the radome away from the direction of the incident field, the effect of the chirality cannot be predicted by a simple theory. Numerical experimentation is needed.

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