

# Effect of Sparse Array Geometry on Estimation of Co-array Signal Subspace

Mehmet Can Hücümenoğlu and Piya Pal  
University of California, San Diego

**Abstract**—This paper considers the effect of sparse array geometry on the co-array signal subspace estimation error for Direction-of-Arrival (DOA) estimation. The second largest singular value of the signal covariance matrix plays an important role in controlling the distance between the true subspace and its estimate. For a special case of two closely-spaced sources impinging on the array, we explicitly compute the second largest singular value of the signal covariance matrix and show that it can be significantly larger for a nested array when compared against a uniform linear array with same number of sensors.

**Index Terms**—Davis Kahan, Difference co-arrays, Sparse arrays, Subspace Estimation.

## I. INTRODUCTION

Sparse array geometries have recently gained significant research interest [1]–[4] owing to several attractive benefits over traditionally used uniform linear arrays (ULA), such as the ability to identify  $O(P^2)$  uncorrelated sources using only  $P$  sensors [1], [2], as well as lower Cramér-Rao bounds and higher spatial resolution than ULAs with the same number of spatial and temporal measurements [4], [5]. These enhanced abilities are attributed to the fact that the difference co-arrays of these sparse arrays contain a ULA segment (consisting of consecutive lags around 0) of size  $O(P^2)$  that can be exploited by algorithms such as co-array MUSIC [1], [4], [6] to resolve more sources than sensors. However, the performance of co-array based DOA estimation algorithms can potentially deteriorate when the so-called signal subspace is not identified properly. In this paper, we study how the geometry of sparse arrays can influence the distance between the true signal subspace and its estimate for the special case of two narrowband sources. We show that the second largest singular value of the signal covariance matrix controls the mismatch between the true and estimated subspaces. In general, greater the second largest singular value, smaller the mismatch. Given a sparse array and ULA with same number of sensors, we show that the second largest singular value of the sparse can be significantly larger than that of the ULA. We will use this result in future to understand non asymptotic performance of sparse arrays with closely spaced sources.

## II. EFFECT OF SPARSE ARRAY GEOMETRY ON SIGNAL SUBSPACE PERTURBATION

Consider a linear array of  $P$  antennas whose physical locations are given by  $\{d_p \lambda/2, p = 1, 2, \dots, P\}$ , where  $d_p$  belongs to an integer set  $\mathbb{S}$ ,  $|\mathbb{S}| = P$ , and  $\lambda$  is the wavelength of far-field narrowband sources impinging on the array. In this

paper we consider the special case of two narrowband sources with Direction-of-Arrival (DOA)  $\theta_1, \theta_2 \in (-\pi/2, \pi/2]$ . The data vector  $\mathbf{y} \in \mathbb{C}^P$  consisting of measurements collected at the  $P$  antennas is given by:

$$\mathbf{y} = \mathbf{A}\mathbf{x} + \mathbf{n}. \quad (1)$$

Here  $\mathbf{A} = [\mathbf{a}(\theta_1) \ \mathbf{a}(\theta_2)]$  is the array manifold matrix where  $\mathbf{a}(\theta_i) = [e^{j\pi d_1 \sin(\theta_i)} \ \dots \ e^{j\pi d_P \sin(\theta_i)}]^T$  represents the array steering vector corresponding to the direction  $\theta_i$ . Furthermore  $\mathbf{x} = [x_1, x_2]^T$  is the vector of complex amplitude of the two sources and  $\mathbf{n}$  represents the additive noise at the  $P$  antennas. **Role of Difference Co-array:** If we assume that the source amplitudes  $x_1, x_2$  are zero-mean statistically uncorrelated random variables, and the noise is zero-mean white and statistically independent of  $\mathbf{x}$ , the correlation matrix of  $\mathbf{y}$  is given by:

$$\mathbf{R}_{\mathbf{y}\mathbf{y}} = \mathbf{A}\mathbf{P}\mathbf{A}^H + \sigma_n^2 \mathbf{I}_P, \quad (2)$$

where  $\mathbf{P} = \text{diag}(p_1, p_2)$  is a diagonal matrix with  $p_1 = \mathbb{E}(|x_1|^2)$ ,  $p_2 = \mathbb{E}(|x_2|^2)$  and  $\mathbb{E}(\mathbf{n}\mathbf{n}^H) = \sigma_n^2 \mathbf{I}$ . Notice that the cross-correlation between the measurement at the  $m$ th and  $k$ th antennas is given by:

$$[\mathbf{R}_{\mathbf{y}\mathbf{y}}]_{m,k} = \sum_{i=1}^2 e^{j\pi \sin \theta_i (d_m - d_k)} p_i + \sigma_n^2 \delta[m - k]. \quad (3)$$

In other words, the cross-correlation depends on the pairwise difference  $\{d_m - d_k\}$  between the sensor locations. This naturally leads to the notion of a difference set [1].

**Definition II.1.** The difference set of a set of integers  $\mathbb{S}$  is defined as  $\mathbb{D}_{\mathbb{S}} = \{d_i - d_j | d_i, d_j \in \mathbb{S}\}$ .

Given a sensor array  $\mathbb{S}$ , we can associate a “virtual difference co-array” whose element locations are given by the set  $\mathbb{D}_{\mathbb{S}}$ . We can also define an integer  $N_{\max}$  as the largest integer such that  $\{0, 1, \dots, N_{\max}\} \subset \mathbb{D}_{\mathbb{S}}$ . In this case, the set  $\mathbb{U}_{\mathbb{S}} = \{0, 1, \dots, N_{\max}\}$  is called the “Non-negative ULA segment” of the difference co-array. We can extract  $N_{\max} + 1$  entries  $\mathbf{R}_{m,n}$  of the data correlation matrix such that  $\{d_m - d_n\} \in \mathbb{U}_{\mathbb{S}}$  and collect them in a vector  $\mathbf{z} \in \mathbb{C}^{N_{\max}+1}$  [1]. Using (3), it can be shown that:

$$\mathbf{z} = \mathbf{A}_{\mathbb{U}_{\mathbb{S}}}\mathbf{p} + \sigma_n^2 \mathbf{e}, \quad (4)$$

where  $\mathbf{A}_{\mathbb{U}_{\mathbb{S}}} \in \mathbb{C}^{(N_{\max}+1) \times 2}$  is the array manifold matrix corresponding to a ULA with element locations given by the set  $\mathbb{U}_{\mathbb{S}}$  and  $\mathbf{e} = [1, 0, 0, \dots, 0]$  is a canonical basis vector in  $\mathbb{C}^{N_{\max}+1}$ . It is well-known that for specially designed sparse arrays such as nested and coprime arrays [1], [2],  $N_{\max} = \Theta(P^2)$  whereas for ULA,  $N_{\max} = \Theta(P)$ . The large difference set of sparse arrays can be utilized to localize more sources than sensors using the so-called co-array MUSIC

Work supported in parts by NSF CAREER Award 1700506, ONR N00014-19-1-2227 and the University of California, San Diego.

algorithm [1], [4], [6]. In this paper, we investigate how the large difference co-array of nested arrays helps to characterize subspace estimation errors.

#### Distance between Perturbed and True Signal Subspaces:

Let  $\hat{\mathbf{z}}$  be an estimate of  $\mathbf{z}$  obtained from the sample covariance matrix  $\hat{\mathbf{R}}_{yy}$ . Then, we have:

$$\text{Toep}(\hat{\mathbf{z}}) = \mathbf{A}_{U_S} \mathbf{P} \mathbf{A}_{U_S}^H + \mathbf{H}, \quad (5)$$

where  $\text{Toep}(\hat{\mathbf{z}})$  represents a Hermitian Toeplitz matrix whose first column is  $\hat{\mathbf{z}}$  and  $\mathbf{H} = \text{Toep}(\sigma_n^2 \mathbf{e} + \hat{\mathbf{z}} - \mathbf{z})$  captures the effect of noise and estimation error. The co-array signal subspace is defined as the span of the co-array steering vectors corresponding to the source directions  $\theta_1$  and  $\theta_2$  and is given by [4]:

$$\mathcal{S}_{ca} = \text{Range}(\mathbf{A}_{U_S}). \quad (6)$$

Algorithms such as co-array MUSIC [1], [4] aim to obtain an estimate of this subspace, which is subsequently used for identifying the DOAs using similar principle as the classical MUSIC algorithm [7]. A popular practice is to use the following estimate of  $\mathcal{S}_{ca}$ :

$$\hat{\mathcal{S}}_{ca} = \text{Range}(\hat{\mathbf{U}}), \quad (7)$$

where the columns of  $\hat{\mathbf{U}} \in \mathbb{C}^{(N_{\max}+1) \times 2}$  are the singular vectors of  $\text{Toep}(\hat{\mathbf{z}})$  corresponding to the two largest singular values. It is obvious that the accuracy of DOA estimation crucially depends on the subspace estimation error, which in turn, depends on the array geometry. Let  $\mathbf{U}$  be an orthogonal basis for  $\mathcal{S}_{ca}$ . Let  $\sigma_2$  be the second largest singular value of  $\mathbf{A}_{U_S} \mathbf{P} \mathbf{A}_{U_S}^H$ . If  $\sigma_2 > \|\mathbf{H}\|_2$ , we can use Davis-Kahan theorem [8] to bound the distance between the true subspace and its estimate as:

$$\text{dist}(\mathbf{U}, \hat{\mathbf{U}}) = \|\mathbf{U}\mathbf{U}^H - \hat{\mathbf{U}}\hat{\mathbf{U}}^H\|_2 \leq \frac{\|\mathbf{H}\|_2}{\sigma_2 - \|\mathbf{H}\|_2}. \quad (8)$$

Here,  $\|\mathbf{H}\|_2$  is the spectral norm of the matrix  $\mathbf{H}$ . The condition  $\sigma_2 > \|\mathbf{H}\|_2$  is also related to the assumption of “no subspace swap”, which is crucial for analysis of co-array MUSIC [4]. We next explicitly characterize the role of co-array geometry in determining how large  $\sigma_2$  can be by considering two array geometries: ULA and nested array. Let  $L = N_{\max} + 1$  and  $\beta = (\sin(\theta_1) - \sin(\theta_2))/2$  is the normalized angle distance. Using the fact that the smallest eigenvalue of  $\mathbf{A}_{U_S}^H \mathbf{P} \mathbf{A}_{U_S}$  is equal to the second largest singular value of  $\mathbf{A}_{U_S} \mathbf{P} \mathbf{A}_{U_S}^H$ , we explicitly compute the second largest singular value of  $\mathbf{A}_{U_S} \mathbf{P} \mathbf{A}_{U_S}^H$  as:

$$\sigma_2 = \frac{L(p_1 + p_2)}{2} - \sqrt{\left(\frac{L(p_1 - p_2)}{2}\right)^2 + p_1 p_2 \frac{1 - \cos(2\pi L\beta)}{1 - \cos(2\pi\beta)}}. \quad (9)$$

Notice that  $\sigma_2$  grows linearly with  $L = N_{\max} + 1$ , when other quantities are held constant. Recall that for a ULA,  $N_{\max} = P - 1$  and for a nested array,  $N_{\max} = \lfloor \frac{P}{2} \rfloor (\lfloor \frac{P}{2} \rfloor + 1)$  [1]. Therefore, for a nested array,  $\sigma_2$  grows quadratically with the number of sensors  $P$  whereas for a ULA, it only grows linearly with  $P$ . This is illustrated in Fig. 1. In Fig. 2, we plot  $\sigma_2$  as a function of  $\beta$  which is the normalized separation between source directions. For most values of the separation between sources,  $\sigma_2$  for the nested array is significantly larger than that of a ULA.

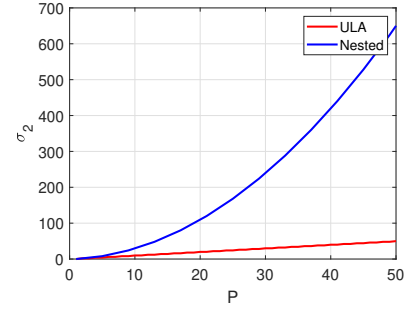


Fig. 1. Dependence of  $\sigma_2$  on the number of antennas  $P$  for nested array and ULA. Here,  $p_1 = p_2 = 1$  and  $\beta = \frac{1}{2}$ .

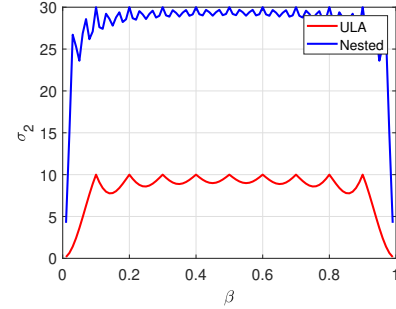


Fig. 2. Dependence of  $\sigma_2$  on  $\beta$  for nested array and ULA. Here,  $P = 10$  for both arrays and  $p_1 = p_2 = 1$ .

### III. CONCLUSION

We studied the effect of sparse array geometry on co-array signal subspace estimation. For the special case of two sources, we explicitly characterized the second largest singular value and how it controls the distance between the true signal subspace and any estimate. In future, we will use these results to characterize non-asymptotic performance of sparse array based DOA estimation algorithms.

### REFERENCES

- [1] P. Pal and P. Vaidyanathan, “Nested arrays: A novel approach to array processing with enhanced degrees of freedom,” *IEEE Transactions on Signal Processing*, vol. 58, no. 8, pp. 4167–4181, 2010.
- [2] P. P. Vaidyanathan and P. Pal, “Sparse sensing with co-prime samplers and arrays,” *IEEE Transactions on Signal Processing*, vol. 59, no. 2, pp. 573–586, 2011.
- [3] H. Qiao and P. Pal, “On maximum-likelihood methods for localizing more sources than sensors,” *IEEE Signal Processing Letters*, vol. 24, no. 5, pp. 703–706, 2017.
- [4] M. Wang and A. Nehorai, “Coarrays, music, and the cramer–rao bound,” *IEEE Transactions on Signal Processing*, vol. 65, no. 4, pp. 933–946, Feb. 2017.
- [5] A. Koochakzadeh and P. Pal, “Cramer–rao bounds for underdetermined source localization,” *IEEE Signal Processing Letters*, vol. 23, no. 7, pp. 919–923, 2016.
- [6] P. Pal and P. P. Vaidyanathan, “Coprime sampling and the music algorithm,” in *2011 Digital Signal Processing and Signal Processing Education Meeting (DSP/SPE)*. IEEE, 2011, pp. 289–294.
- [7] R. Schmidt, “Multiple emitter location and signal parameter estimation,” *IEEE Transactions on Antennas and Propagation*, vol. 34, no. 3, pp. 276–280, 1986.
- [8] C. Davis and W. M. Kahan, “The rotation of eigenvectors by a perturbation. iii,” *SIAM Journal on Numerical Analysis*, vol. 7, no. 1, pp. 1–46, 1970.