Design of Wideband Planar Absorbers using Composite Materials

¹E. A. Hashish, ¹S. M. Eid, and ²S. F. Mahmoud

¹Department of Electronics and Communication, Faculty of Engineering, Cairo University, Egypt essamhh@ieee.org

² Electrical Engineering Department, Kuwait University, P.O. Box 5969, Safat 13060, Kuwait Samirfm2000@yahoo.com

Abstract – Design of planar microwave absorbers usually incorporates the use of either magnetic or electric lossy materials. In this study, chiral materials are included in the design process of these absorbers. The genetic algorithm is used to obtain suitable solutions satisfying the design requirements. Wideband absorbers are designed using different configurations and presented in this paper. It is shown that the inclusion of chiral materials in the design process leads to more efficient absorbers.

Keywords: Electromagnetic absorbers, chiral materials, composite materials.

I. INTRODUCTION

Wideband microwave absorbers are of great interest for their important applications. These applications include radar cross section reduction of a wide range of objects, suppression of unwanted radiation and development of anechoic chambers [1-3]. Different designs of such absorbers are presented for various configurations including single layer [4-8], two layers [9-12] and multilayer absorbers [3, 13-16]. The design techniques of these absorbers are based on graphical methods [4, 7], local optimization methods [13], global optimization methods [3, 14, and 15] and analytical methods [9-11]. All of the above designs utilize layers of absorber materials which are of simple lossy electric or lossy magnetic types.

In this paper, chiral materials are used in the design process in addition to the above materials to construct the absorber. Design of wideband absorbers with different layer configurations is performed with the aid of the genetic algorithm as a global optimization technique. Better performance is expected when chiral materials are included in the design procedure of these absorbers, by virtue of the extra degree of freedom provided by the chirality parameter.

II. FORMULATION OF THE PROBLEM

Consider a planar absorber that is composed of N layers of lossy materials backed by a perfectly conducting

surface as shown in Fig. 1. Each layer is defined by its complex permittivity ε_n , complex permeability μ_n , thickness d_n , in addition to chiral admittance ζ_n where $I \leq n \leq N$. The time variation of the electromagnetic fields is assumed sinusoidal with the factor $e^{j\omega t}$. The electromagnetic fields inside any layer is controlled by the equations [16],

$$\underline{D}_n = \varepsilon_n \, \underline{E}_n - j \, \xi_n \, \underline{B}_n \tag{1}$$

$$\underline{B}_n = \mu_n \left(\underline{H}_n + j \,\xi_n \,\underline{E} \right),\tag{2}$$

where ε_n, μ_n are the usual electric permittivity and magnetic permeability, and ξ_n is the chiral admittance.

Let a uniform plane wave be normally incident to the absorber interface with the air.

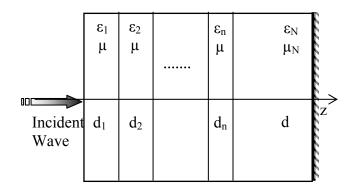


Fig. 1. N-layer planar absorber.

Because of the chirality, there are two normal propagating modes in each layer; one is having right circular polarization (RCP) with $\exp(-jk_n^+z)$ propagation factor and the other left circular polarization (LCP) with $\exp(-jk_n^-z)$ propagation factor where [17,18],

$$k_n^{\pm} = \sqrt{\omega^2 \mu_n \varepsilon_n + \omega^2 \mu_n^2 \xi_n^2} \pm \omega \mu_n \xi_n \qquad (3)$$

writing ξ_n in the form,

$$\xi_n = \sqrt{\frac{\varepsilon_0}{\mu_0}} \,\chi_n \tag{4}$$

where χ_n is a dimensionless chirality factor, we re-express k_n^{\pm} as,

$$k_n^{\pm} = k_n \left[\sqrt{1 + \left(\frac{\eta_n}{\eta_0}\right)^2 \chi_n^2} \pm \frac{\eta_n}{\eta_0} \chi_n \right]$$
(5)

where

$$k_n = \omega \sqrt{\mu_n \,\varepsilon_n} \tag{6}$$

and

$$\eta_n = \sqrt{(\mu_n / \varepsilon_n)} , \qquad (7)$$

with η_0 is the free space wave impedance.

It can be shown that the reflection coefficient Γ of a circularly polarized plane wave, which is normally incident on the absorber, is given by,

$$\Gamma = \frac{z_1 - \eta_0}{z_1 + \eta_0} \tag{8}$$

where,

Z

$$_{n} = \zeta_{n} \frac{\zeta_{n} \tanh(\gamma_{n} d_{n}) + z_{n+1}}{\zeta_{n} + z_{n+1} \tanh(\gamma_{n} d_{n})}, \qquad (9)$$

with

and

$$\zeta_{n} = \frac{\eta_{n}}{\sqrt{1 + (\eta_{n} / \eta_{0})^{2} \chi^{2}}}, \qquad (10)$$

$$z_N = \zeta_n \tanh(\gamma_n d_n), \qquad (11)$$

$$\gamma_n = j \, \frac{\left(k_n^+ + k_n^-\right)}{2} = j k_n \,,$$
 (12)

where γ_n is the average propagation constant of the forward and backward plane waves in the chiral medium, which have opposite circular polarization. We note here that equation (8) applies to either an RCP or LCP incident wave. Hence it is also valid for a linearly polarized incident wave. On the other hand, the case when $\chi_n = 0$ represents either an ordinary lossy electric or lossy magnetic material, with no chirality.

It is worth noting that in the design process of chiral absorbers, only magnetic materials are considered as chiral materials while electric materials are still nonchiral. In other words, the chirality is imposed on magnetic materials only and this is quite sufficient to achieve good performance for the designed absorbers. Moreover, all the materials that are used to construct the absorbers are assumed to be dispersive. This dispersion is imposed such that the complex relative permittivity of the electric layers and the complex relative permeability of the magnetic layers are inversely proportional to square root of the operating frequency. In such case, the electric and magnetic parameters of these layers \mathcal{E}_n and μ_n are expressed as,

$$\varepsilon_n = \varepsilon_0 \,\varepsilon_m \,(1 - \tan \delta_{en}) \,\sqrt{f} \,/f_0 \qquad (13)$$
$$\mu_n = \mu_0$$

for the electric layers, and,

$$\varepsilon_n = 10\varepsilon_0$$

$$\mu_n = \mu \,\mu_m \left(1 - \tan \delta_{mn}\right) \sqrt{f / f_0}$$
(14)

for the magnetic layers. This frequency dependence is satisfied by most materials.

The depth of each layer d_n is normalized with respect to the wavelength of the wave inside this layer.

III. APPLICATION OF THE GENETIC ALGORITHM

A conventional genetic algorithm is built to solve the optimization problem, whose objective is to minimize the total reflected power from the layered structure, over a wide frequency range. We choose to maximize the total power transmitted to the absorber and at the same time minimize the maximum power reflected over the frequency band. A set of constraints is used to impose limitations or specifications on the system parameters. The variables (genes) in the fitness function are the intrinsic electric and magnetic parameters, the chirality factor as well as the thickness of the layers.

The genetic algorithm is applied for 100 design experiments in each layer configuration of chiral and non chiral absorbers with 6000 iterations in each experiment. In each design experiment, the fitness criterion is examined over a normalized frequency range from 0.1 to 9.0 with five samples in the range. Of course, larger number of samples within the range can be considered; however, this would cost more computational time without guarantee of much better results. The fitness criterion is defined as,

$$fitness = \frac{1}{2} \mathcal{Q} \left\{ \sum_{i=1}^{N_s} (1 - \left| \Gamma_i^2 \right|) \right\}$$

$$+ (1 - \mathcal{Q}) \left\{ 1 - \left| \Gamma_{\max}^2 \right| \right\}$$

$$(15)$$

where Q is a factor ranging from 0 to 1 and it is set to 0.5 in these experiments, N_s , is the number of applied samples through the frequency range, Γ_i is the reflection coefficient at any sampling frequency, Γ_{max} is the maximum reflection coefficient amplitude among the samples.

IV. RESULTS

The best four designs in the above experiments are chosen for each layer configuration of the absorber based on the fitness criterion equation (15). The parameters of these designs are given in Tables 1 to 3 for two, three and four layer chiral absorber models. Examination of the fitness criteria in these leads to the conclusion that the absorber performance improves with the increased number of layers. The frequency responses of these chiral absorbers are presented in Figs. 2 to 4. The best design of each layer configuration exhibits reflection level below than -20 dB all over the frequency range. This level decreases obviously as the number of layers increases. The frequency response of the best four design experiments of the five layer nonchiral absorber is shown in Fig. 5. It is clear that the frequency response of these designs has much lower performance than those of the chiral absorbers.

Table 1. Design parameters for a two layer chiral absorber.

| | Layer No. | Type | $\epsilon_{\rm rn}$, $\tan(\delta_n/2)$ | μ_{rn} , Tan ($\delta_{n}/2$) | dn/λ n | Chirality |
|----------------------------------|-----------|------|--|---|--------|-----------|
| Design 1 Fitness = 99.518% | 1 | mag. | 10.0, 0.0 | 7.48, .75 | 099 | .779 |
| | 2 | mag. | 10.0, 0.0 | 29.8, .797 | 076 | .556 |
| Design 2 Fitness = 99.332% | 1 | mag. | 10.0, 0.0 | 5.86, .75 | .099 | .691 |
| | 2 | mag. | 10.0, 0.0 | 35.1, .797 | .075 | .622 |
| Design 3 Fitness = 99.328% | 1 | mag. | 10.0, 0.0 | 10.08, .65 | .08 | .772 |
| | 2 | mag. | 10.0, 0.0 | 25.0, .80 | .09 | .694 |
| Design 4 Fitness = 99.202% | 1 | mag. | 10.0, 0.0 | 20.36, .737 | .07 | .783 |
| | 2 | mag. | 10.0, 0.0 | 36.1, .787 | .06 | .516 |

A comparison between the chiral and nonchiral absorbers according to the best fitness criterion is given in Fig. 6 for each layer configuration. This comparison shows clearly that the chiral absorbers have much better performance than the nonchiral one. Even, it is evident that the fitness criterion of the two layer chiral absorber is better than the five layer nonchiral one. The conclusion is that less number of chiral layers is needed to achieve a prescribed reflection level over a wide frequency bandwidth.

Table 2. Design parameters for a three layer chiral absorber.

| | Layer No. | Type | $\epsilon_{\rm rn},$ tan ($\delta_{\rm n}/2$) | μ_{rn} tan ($\delta_n/2$) | d ո/Ղ ո | Chirality |
|----------------------------------|-----------|-------|---|------------------------------------|----------------|-----------|
| Design 1 Fitness = 99.842% | 1 | mag. | 10.0, 0.0 | 18.04 .7148 | .05311 | .746 |
| | 2 | elec. | 9.353, .065 | 1.0, 0.0 | .03647 | .000 |
| | 3 | mag. | 10.0, 0.0 | 26.76, .792 | .09839 | .541 |
| Design 2 Fitness = 99.814% | 1 | mag. | 10.0, 0.0 | 18.46, .552 | .08656 | .78 |
| | 2 | elec. | 3.93, .629 | 1.0, 0.0 | .04328 | .00 |
| | 3 | mag | 10.0, 0.0 | 36.53, .796 | .08245 | .36 |
| Design 3 Fitness = 99.808% | 1 | mag | 10.0, 0.0 | 22.62, .790 | .02281 | .78 |
| | 2 | elec. | 9.41, .713 | 1.0, 0.0 | .01375 | .00 |
| | 3 | mag | 10.0, 0.0 | 38.3, .799 | .09689 | .44 |
| Design 4 Fitness = 99.75% | 1 | mag | 10.0, 0.0 | 23.7, .769 | .03856 | .80 |
| | 2 | elec. | 8.89, .054 | 1.0, 0.0 | .02695 | .00 |
| | 3 | mag | 10.0, 0.0 | 36.6, .798 | .08868 | .62 |

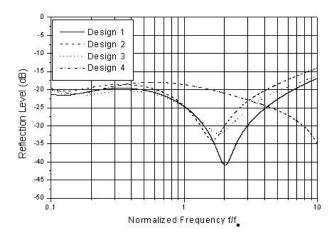


Fig. 2. Power reflection level for a two-layer chiral model.

| | Layer No. | Type | $^{\rm E}_{\rm rm}$ tan ($\delta_{\rm n}/2$) | $\mu_{\rm rn}$, tan ($\delta_{\rm n}/2$) | $\mathbf{d}_{n}/\boldsymbol{\lambda}_{n}$ | Chirality |
|----------------------------------|-----------|-------|--|---|---|-----------|
| Design 1 Fitness = 99.97% | 1 | elec. | 3.43, .448 | 1.0, 0.0 | .001 | .00 |
| | 2 | mag. | 10.0, 0.0 | 14.39, .781 | .04 | .80 |
| | 3 | elec. | 7.50, .198 | 1.0, 0.0 | .044 | .00 |
| | 4 | mag. | 10.0, 0.0 | 26.16, .793 | .097 | .45 |
| Design 2 Fitness = 99.99 % | 1 | mag. | 10.0, 0.0 | 8.50, .707 | .055 | .70 |
| | 2 | elec. | 3.21, .422 | 1.0, 0.0 | .056 | .00 |
| | 3 | mag. | 10.0, 0.0 | 37.2, .66 | .022 | .49 |
| | 4 | mag | 10.0, 0.0 | 32.4, .793 | .092 | .64 |
| | 1 | mag. | 10.0, 0.0 | 24.0, .666 | .054 | .77 |
| Design 3 Fitness = | 2 | elec. | 6.15, .387 | 1.0, 0.0 | .025 | .00 |
| 99.872% | 3 | mag. | 10.0, 0.0 | 30.1, .795 | .036 | .58 |
| | 4 | Mag | 10.0, 0.0 | 39.79, .791 | .089 | .79 |
| | 1 | Mag | 10.0, 0.0 | 30.40, .486 | .089 | .77 |
| Design 4 Fitness = | 2 | Mag | 10.0, 0.0 | 25.68, .746 | .068 | .77 |
| 99.864% | 3 | elec. | 7.15, .519 | 1.0, 0.0 | .097 | .00 |
| | 4 | Mag | 10.0, 0.0 | 27.59, .784 | .098 | .49 |

Table 3. Design parameters for a four layer chiral absorber.

To simply explain why chiral absorbers surpass the non chiral ones, refer to the layer intrinsic impedance parameter in equation (10).

When the layer is magnetic with high loss, such that $|\mu/\mu_0|\chi^2 >> |\varepsilon/\varepsilon_0|$, then $\zeta \approx \eta_0/\chi$. So a highly lossy magnetic layer with chirality parameter $\chi = 1$, is a good match to air! (over a wide band) [19]. Of course $\chi=1$ is too much chirality, so it is hard to manufacture. But this shows that adding one or more layers to the one magnetic layer should lead to a practical absorber. This should explain why few numbers of chiral layers could make a wide-band good absorber.

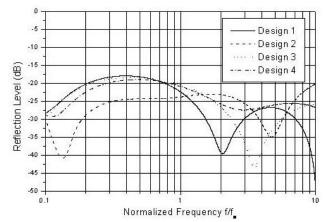


Fig. 3. Power reflection level for a three-layer chiral model.

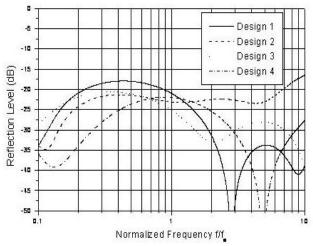


Fig. 4. Power reflection level for a four-layer chiral model.

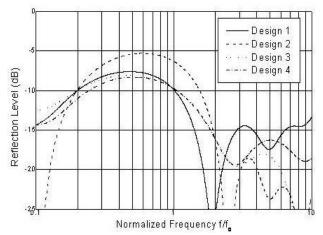


Fig. 5. Power reflection level for a five-layer non-chiral model.

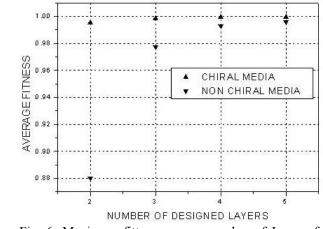


Fig. 6. Maximum fittness versus number of Layers for chiral and non-chiral media.

V. CONCLUSION

In this paper, the analysis of chiral planar absorber is presented. The genetic algorithm is applied to obtain the best four designs over 100 design experiments for each layer configuration of chiral and non-chiral absorbers. Inspection of the fitness criterion of these experiments indicates that the chiral absorbers have much better performance than non-chiral ones. It is concluded that much less number of layers is needed to achieve a prescribed maximum level of the reflection coefficient over a wide frequency range.

REFERENCES

- [1] J. A. Adam, "How to design an invisible aircraft," *IEEE Spectrum*, vol. 25, pp. 26-31, April 1988.
- [2] D. S. Weile, E. Michielssen, and D. E. Goldberg, "Genetic algorithm design of Pareto optimal broadband microwave absorbers," *IEEE Trans. EM Compatibility*, vol. 38, no. 3, pp. 518-525, August 1996.
- [3] S. F. Mahmoud, "Design of a planar microwave absorber," *ICECS'97, Cairo, Egypt*, pp. 1110-1113, December 1997.
- [4] H. M. Musal, J. R. and H. T. Hahn, "Thin-layer electromagnetic absorber design," *IEEE Trans. Mag.* vol. 25, no. 5, pp. 3851-3853, September 1989.
- [5] R. L. Fante and M. T. Mc Cromick, "Reflection properties of the Salisbury screen," *IEEE Trans. AP*, vol. 36, no. 10, pp. 1443-1454, October 1988.
- [6] E. F. Knott, "The thickness criterion for single-layer radar absorbents," *IEEE AP*, vol. 27, pp. 698-701, 1979.
- [7] F. A. Frenandez and A. Q. Valenzuela, "General solution for single-layer electromagnetic wave absorber," *Electronics Letters*, vol. 21, no. 1, pp. 20-21, January 1985.

- [8] A. Q. Valenzuela and F. A. Frenandez, "General design theory for single-layer homogeneous absorber," *IEEE Trans. AP.*, vol. 44, no. 7, pp. 822-826, July 1996.
- [9] S. F. Mahmoud, "A two-layer planar microwave absorber," *Microwave and Optical Technology Letters*, vol. 15, no. 3, pp. 170-173, June 1997.
- [10] S. F. Mahmoud and M. K. Habib, "Design of a twolayer microwave absorber," *Journal of Electromagnetic waves and Applications*, vol. 12, pp. 1005-1014, 1998.
- [11] E. A. Hashish, "Design of wideband thin layer planar absorber," *Electromagnetic Waves and Applications*, vol. 16, no. 2, pp. 227-241, 2002.
- [12] B. Chambers, "Frequency tuning characteristics of an adaptive Jaumann radar absorber incorporating variable impedance layers," *Electronics Letters*, vol. 30, no. 22, pp. 1892-1893, October 1994.
- [13] J. J. Pesque, D. P. Bouche, and R. Mittra, "Optimization of multilayer antireflection coatings using an optimal control method," *IEEE Trans. MTT*, vol. 40, no. 9, pp. 1789-1796, September 1992.
- [14] E. Michielssen, J. M. Sajer, S. Ranjithan, and R. Mittra, "Design of lightweight, broad-band microwave absorbers using genetic algorithms," *IEEE Trans. MTT.*, vol. 41, no. 6/7, pp. 1024-1030, June/July 1993.
- [15] D. G. Li and A. C. Watson "Optical thin film optimization design using genetic algorithms," *IEEE International Conference on intelligent processing* systems, Oct. 28-31 Bejing, China, pp. 132-136, 1997.
- [16] C. N. Chiu and I. T. Chiang, "Transient reflection properties of a dispersive and lossy Bi-isotropic slab with an anisotropic laminated composite packing," *IEEE Trans. On EMC*, vol. 47, no. 4, pp. 845-852, November 2005.
- [17] N. Engheta and D. L. Jaggard, "Electromagnetic chirality and its applications," *IEEE Antennas Propagat. Soc. Newsletter*, vol. 30, no. 5, pp. 6-12, 1988.
- [18] S. F. Mahmoud, "Mode characteristics in chirowaveguides with constant impedance walls," *Journal of Electromagnetic Waves and Applications*, (JEMWA), vol. 6, no. 5/6, pp. 625-640, 1992.
- [19] J. C. Liu and D. L. Jaggard, "Chiral layers on planar surfaces," *Journal of Electromagnetic Waves and Applications, (JEMWA)*, vol. 6, no. 5/6, pp. 651-667, 1992.



Samir F. Mahmoud graduated from the Electronic Engineering Dept., Cairo university, Egypt in 1964. He received the M.Sc and Ph.D. degrees in the Electrical Engineering Department, Queen's university, Kingston, Ontario, Canada in 1970 and 1973. During the academic year 1973-1974, he was a visiting research fellow at the Cooperative Institute for Research

in Environmental Sciences (CIRES). Boulder, CO, doing research on Communication in Tunnels. He spent two sabbatical years, 1980-1982, between Queen Mary

College, London and the British Aerospace, Stevenage, where he was involved in design of antennas for satellite communication. Currently Dr. Mahmoud is a full professor at the EE Department, Kuwait University. Recently, he has visited several places including Interuniversity Micro-Electronics Centre (IMEC), Leuven, Belgium and spent a sabbatical leave at Queen's University and the royal Military College, Kingston, Ontario, Canada in 2001-2002. His research activities have been in the areas of antennas, geophysics, tunnel communication, e.m wave interaction with composite materials and microwave integrated circuits. Dr. Mahmoud is a Fellow of IET and one of the recipients of the best IEEE/MTT paper for 2003.