Enhanced Bandwidth Small Square Slot Antenna with Circular Polarization Characteristics for WLAN/WiMAX and C-Band Applications

Mohammad Ojaroudi¹, Nasser Ojaroudi², and Noradin Ghadimi¹

¹ Young Researchers Club Ardabil Branch, Islamic Azad University, Ardabil, Iran m.ojaroudi@iauardabil.ac.ir and noradin.ghadimi@gmail.com

² Department of Electrical Engineering Ardabil Branch, Islamic Azad University, Ardabil, Iran n ojaroudi@srttu.edu

Abstract - This article proposes a novel printed with circular polarization slot antenna characteristics using an L-shaped slot in the ground plane and an L-shaped radiating stub with a pair of Γ -shaped slits for simultaneously satisfying wireless local area network (WLAN), worldwide interoperability for microwave access (WiMAX), and C-Band applications. The operating frequencies of the proposed antenna are 5.2/5.8 GHz, which covers WLAN system, 5.5 GHz for WiMAX system, and 4 GHz for C-Band system. The desired resonant frequencies are obtained by adjusting the length of Γ -shaped slits in the both sides of tapered microstrip feed-line. The measured results show good agreement with the numerical prediction. The designed antenna has a small size of $20 \times 20 \text{ mm}^2$.

Index Terms — Circular polarization performance, Γ -shaped slit, and printed square slot antenna.

I. INTRODUCTION

Circular polarization is one of the common polarization schemes used in current wireless communication systems, such as radar and satellite systems. This is due to the fact that it can provide better mobility and weather penetration than linear polarization microstrip patch antenna [1], spiral antenna [2], dielectric resonator antenna [3], and slot antenna [4], which are some of the typical types of circularly polarized (CP) antennas. The operation principle of these CP antennas is to excite two orthogonal field components with equal amplitudes.

In this paper, a new printed square slot antenna with single-feed circular polarization characteristics, for WLAN/WiMAX and C-Band applications is presented. First by inserting an Lshaped slot on the ground plane, additional resonances are excited and the bandwidth is improved that achieves a fractional bandwidth with dual resonance performance of more than 45%. In the proposed structure, wide band circular polarization function is provided by cutting an Lshaped slot on the ground plane and a pair of Γ shaped slits in the corners of the L-shaped radiating stub. The size of the designed antenna is smaller than the slot antennas with circular polarization function reported recently [1-4].

II. ANTENNA DESIGN

The proposed square slot antenna fed by a microstrip line is shown in Fig. 1, which is printed on an FR4 substrate of thickness 0.8 mm, and permittivity 4.4. By cutting a novel slot of suitable dimensions at the ground plane, circular polarization can be constructed. The truncated ground plane and L-shaped radiating stub with two Γ -shaped slits are playing an important role in the circular polarization characteristics of this antenna. This is because thev can adjust the electromagnetic coupling effects between the feedline and the ground plane, and improves its impedance bandwidth without any cost of size or expense [4]. Based on electromagnetic coupling, the modified L-shaped radiating stub with a pair of Γ -shaped slits act as an impedance matching element to control the impedance bandwidth of the proposed antenna. Because it can creates additional surface current paths in the antenna and thus additional resonances are excited and much wider impedance bandwidth can be produced [5].

In this work, we start by choosing the aperture length L_s+L_{sl} . We have a lot of flexibility in choosing this parameter. The length of the aperture mostly affects the antenna bandwidth. As L_s+L_{sl} decreases, so does the antenna BW and vice versa. At the next step, we have to determine the aperture width aperture $W_s + W_{sl}$. The width is approximately $\lambda_s/2$, where λ_s is the slot wavelength. The wavelength λ_s depends on a number of parameters such as the slot width as well as the thickness and dielectric constant of the substrate on which the slot is fabricated. The last and final step in the design is to choose the length of the Γ -shaped slits resonators. A good starting point is to choose it to be equal to $\lambda_m/4$, where λ_m is the guided wavelength in the microstrip line. In this design, the optimized length $L_{resonance}$ is set to resonate at $0.25\lambda_{resonance}$, where $L_{resonance1} = W_{L1} +$ $W_{L2} + L_L$ and $L_{resonance2} = W_{L3} + W_{L4} + L_{L1} + L_{f}$. $\lambda_{resonance1}$ and $\lambda_{resonance2}$ correspond to the first resonance frequency (4 GHz) and the second resonance frequency (5.3 GHz), respectively. The final values of presented slot antenna design parameters are specified in Table 1.

Table 1: The final dimensions of the designed antenna.

Param.	mm	Param.	mm	Param.	mm
W _s	8	L_{S}	6	W_{S1}	8
L_{S1}	7	W_{f}	0.5	W_{f1}	2
W _L	4	L_L	8	W_{L1}	3
L_{L1}	2	W_{L2}	4	W_{L3}	2
W_{L4}	2	L_g	2	W_{g}	3

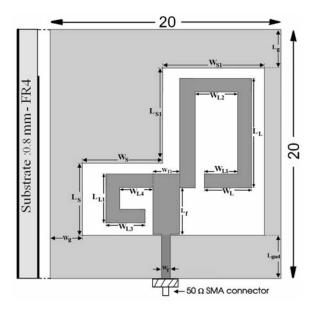


Fig. 1. Geometry of the proposed microstrip-fed slot antenna.

III. RESULTS AND DISCUSSIONS

In this Section, the slot antenna with various design parameters were constructed, and the numerical and experimental results of the input impedance and axial ratio characteristics are presented and discussed. The parameters of this proposed antenna are studied by changing one parameter at a time and fixing the others. The simulated results are obtained using the Ansoft simulation software high-frequency structure simulator (HFSS) [6].

Figure 2 shows the structure of the various antennas used for wide-band impedance and polarization performance simulation circular studies. Return loss and axial ratio characteristics for ordinary square slot antenna with an L-shaped radiating stub and ground plane with a rectangular slot (Fig. 2 (a)), a ground plane with an L-shaped slot (Fig. 2 (b)), and L-shaped radiating stub with a pair of Γ -shaped slits and ground plane with an L-shaped slot (Fig. 2 (c)) are all compared in Fig. 3. As shown in Fig. 3 (a) and (b), in order to generate dual resonance characteristics (4/5.3 GHz), and to improve the circular polarization bandwidth, we use two L-shaped slits in the corners of the radiating stub. Also, by cutting an L-shaped slot in the ground plane a wide band circular polarization function is achieved that cover all the 5.2/5.8 GHz WLAN, 5.5 GHz WiMAX and 4-GHz C band.

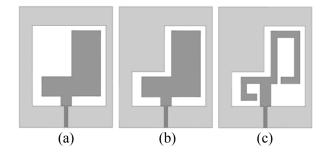


Fig. 2. The ordinary square slot antenna with an L-shaped radiating stub (a) and ground plane with a rectangular slot, (b) and ground plane with an L-shaped slot, (c) with a pair of Γ -shaped slits and ground plane with an L-shaped slot.

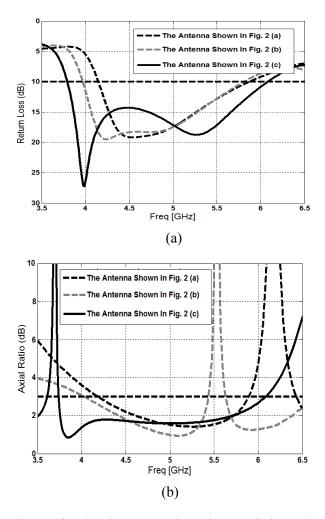


Fig. 3. Simulated (a) return loss characteristics and (b) axial ratio for the antennas shown in Fig. 2.

To understand the phenomenon behind this dual resonances performance, the simulated

current distribution on the radiating stub for the proposed antenna at the new resonance frequencies of 4 GHz and 5.3 GHz are proposed at Fig. 4 (a) and (b), respectively. It can be observed from Fig. 4 (a) and (b) that the current is concentrated at these frequencies on the edges of the interior and exterior of the modified Γ -shaped folded microstrip arms. Therefore, the antenna impedance changes at these frequencies due to the resonance properties of the proposed structure.

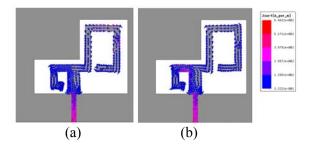


Fig. 4. Simulated surface current distributions for the proposed antenna shown in Fig. 1 at, (a) 4 GHz (first resonance frequency) and (b) 5.3 GHz (second resonance frequency).

In order to investigate the effects of Γ -shaped slits size on the proposed antenna, the simulated return curves with different values of Γ -shaped slits lengths are plotted in Fig. 5. As shown in Fig. 5, when W_L increases from 2 mm to 4.8 mm, the first resonance frequency is varied from 4.3 GHz to 3.8 GHz. From these results, we can conclude that the first resonance frequency is controllable by changing the W_L . Also, as the W_{L3} increases from 1.3 mm to 2.8 mm, the second resonance frequency is varied from 5.1 GHz to 5.5 GHz. Therefore, the second resonance frequency is controllable by changing the W_{L3} .

In this study, to achieve the circular polarization characteristics, we use two Γ -shaped folded microstrip arms in the radiating stub and an L-shaped slot on the ground plane, as displayed in Fig. 1. By changing the dimension of these open stub structures, we can achieve the RHCP and RLCP characteristics. The simulated RHCP and RLCP curves at 4 GHz and 5.3 GHz are presented in Fig. 6 (a) and (b), respectively. These figures confirm that the proposed antenna is capable of supporting both RHCP and RLCP modes at these frequencies. The proposed slot antenna with final design, as shown in Fig. 7, was built and tested.

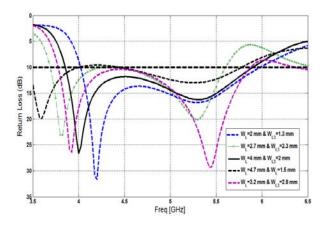


Fig. 5. Simulated return loss characteristics of the proposed antenna with different values of W_L and W_{L3} .

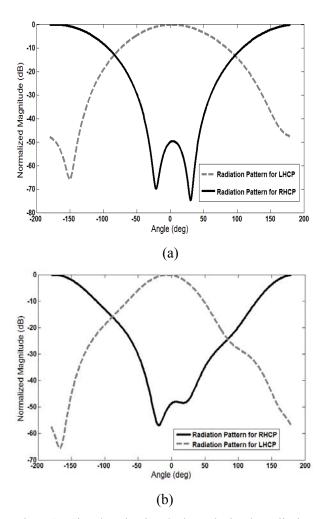


Fig. 6. Simulated circularly-polarized radiation patterns at (a) 4 GHz (first resonance frequency) and (b) 5.3 GHz (second resonance frequency).

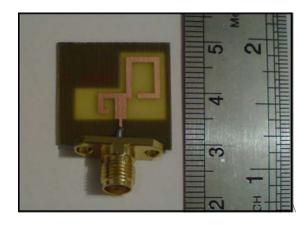
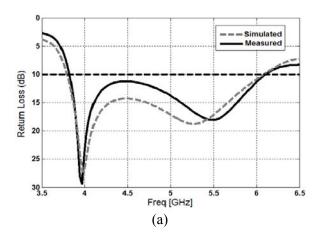


Fig. 7. Photograph of the realized printed slot antenna.

Figure 8 (a) and (b) shows the measured and simulated return loss and axial ratio characteristics of the proposed antenna, respectively. As shown in Fig. 8 (a) and (b), the fabricated antenna has the frequency band of 3.73 GHz to over 6.07 GHz with circular polarization characteristics 3.72 GHz - 6.05 GHz. As shown in Fig. 8, there exists a discrepancy between the measured data and the simulated results. This could be due to the effect of the SMA port. In order to confirm the accurate return loss and axial ratio characteristics for the designed antenna, it is recommended that the manufacturing and measurement process need to be performed carefully.

Figure 9 shows measured maximum gain of the designed antenna in dBi, in the Z-axis direction. A two-antenna technique is used to measure the radiation gain. As shown in Fig. 9, the measured maximum gain has a variation similar to other slot antennas gain at WLAN/ WiMAX and C-band frequencies [7].



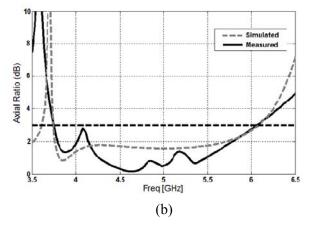


Fig. 8. Measured and simulated (a) return loss characteristics and (b) axial ratio characteristics for the proposed antenna.

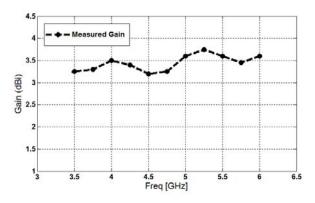


Fig. 9. Measured maximum gain of the proposed antenna.

IV. CONCLUSION

In this paper, a novel square slot antenna with circular polarization characteristics for WLAN/WiMAX and C-band applications is presented. The proposed antenna can operate from 3.73 GHz to 6.07 GHz with circular polarization characteristics around 3.72 GHz - 6.05 GHz. The designed antenna has a small size. Simulated and experimental results show that the proposed antenna could be a good candidate for WLAN/WiMAX and C-band applications.

ACKNOWLEDGMENT

The authors are thankful to Microwave Technology (MWT) Company staff for their beneficial and professional help (www.microwave-technology.com).

REFERENCES

- [1] K. L. Wong, F. S. Chang, and T. W. Chiou, "Lowcost broadband circularly polarized probe-fed patch antenna for WLAN base station," *IEEE Antennas Propag. Soc. Int. Symp.*, vol. 2, pp. 526–529, 2002.
- [2] J. D. Kraus, *Antennas*, 2nd ed. New York: McGraw-Hill, 1988, ch. 7.
- [3] S. Yazdanifard, R. A. Sadeghzadeh, and M. Ojaroudi, "Ultra-wideband small square monopole antenna with variable frequency band notch function," *Progress in Electromagnetics Research C*, vol. 15, pp. 133-144, 2010.
- [4] G. Beigmohammadi, C. Ghobadi, J. Nourinia, and M. Ojaroudi, "Small square slot antenna with circular polarisation characteristics for WLAN/WiMAX applications," *Electronics Letters*, vol. 46, no. 10, pp. 672-673, 2010.
- [5] M. Ojaroudi and A. Faramarzi, "Multi-resonance small square slot antenna for Ultra-wideband applications," *Microwave and Optical Tech. Letters*, vol. 53, no. 9, Sep. 2011.
- [6] Ansoft High Frequency Structure Simulation (HFSS), ver. 13, *Ansoft Corporation*, 2010.
- [7] M. -J. Chiang, T. -F. Hung, and S. -S Bor, "Dualband circular slot antenna design for circularly and linearly polarized operations," *Microwave and Optical Tech. Letters.*, vol. 52, no. 12, pp. 2717-2721, Dec. 2010.



Mohammad Ojaroudi was born on 1984 in Germi, Iran. He received his B.Sc. degree in Electrical Engineering from Azad University, Ardabil Branch and M.Sc. degree in Telecommunication Engineering from Urmia University. From

2010, he has been working toward the Ph.D. degree at Shahid Beheshti University. From 2007 until now, he is a Teaching Assistant with the Department of Electrical Engineering, Islamic Azad University, Ardabil Branch, Iran. Since March 2008, he has been a Research Fellow (Chief Executive Officer) in the microwave technology company (MWT), Tehran, Iran. His research interests include analysis and design of microstrip antennas, design and modeling of microwave structures, radar systems, and electromagnetic theory.



Nasser Ojaroudi was born on 1986 in Germi, Iran. He received his B.Sc. degree in Electrical Engineering from Islamic Azad University, Ardabil Branch. From 2011, he is working toward the M.Sc. degree in Telecommunication Engineering at Shahid Rajaee Teacher Training

University. Since March 2008, he has been a Research Fellow in the Microwave Technology (MWT) Company, Tehran, Iran. His research interests include ultra-wideband (UWB) microstrip antennas and bandpass filters (BPF), reconfigurable structure, design and modeling of microwave device, and electromagnetic wave propagation. He is author and coauthor of more than fifty journal and international conference papers.



Noradin Ghadimi was born in Ardabil-Iran in 1985, and received the B.Sc. degree in electrical engineering from the Islamic Azad University, Ardabil Branch, Ardabil, Iran, in 2009 and the M.Sc. degree in electrical engineering from the Islamic Azad University Ahar Branch, Ahar,

Iran, in 2011. His research interests include Power System Protection, modeling and analysis of Distributed Generations, renewable energy and communications systems.