

# UWB Monopole Antenna with Dual Band-Stop Performance Using G-Shaped SRR and SIR Structures at Feed Line

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**Abstract** — This article presents a new design of ultra-wideband monopole antenna with the characteristics of dual band rejection. The antenna consists of a circular radiating patch, a feed-line with G-shaped Step-Impedance Resonator (SIR) slot and Split-Ring Resonator (SRR) structure, and a ground plane which provides a wide usable fractional bandwidth of more than 120% with two notch bands around 5.15-5.35 GHz and 7.25-7.75 GHz, to avoid interferences from High Performance Radio Local Area Networks (HIPERLAN) and downlink frequency of X-band satellite systems. The simulated and measured results show that the antenna design exhibits an operating bandwidth ( $VSWR < 2$ ) from 2.86 to 12.92 GHz, excluding the rejected bands. The proposed antenna configuration is simple, easy to fabricate and can be integrated into any UWB system. The antenna has an ordinary circular-disc radiating patch, therefore displays good omnidirectional radiation patterns, even at higher frequencies.

**Index Terms** — Dual band-notched function, G-shaped structure, hiperlan, satellite communications, UWB.

## I. INTRODUCTION

After allocation of the frequency band from 3.1 to 10.6 GHz for the commercial use of Ultra-Wideband (UWB) systems by the Federal Communication Commission (FCC) [1], ultra-wideband systems have received phenomenal gravitation in wireless communication. Designing

an antenna to operate in the UWB band is quite a challenge, because it has to satisfy the requirements such as ultra wide impedance bandwidth, omni-directional radiation pattern, constant gain, high radiation efficiency, constant group delay, low profile, easy manufacturing, etc. [2]. In UWB communication systems, one of key issues is the design of a compact antenna while providing wideband characteristic over the whole operating band. Consequently, a number of microstrip antennas with different geometries have been experimentally characterized [3-7].

There are many narrowband communication systems which severely interfere with the UWB communication system, such as the High Performance Radio Local Area Networks (HIPERLAN) operating at 5.15-5.35 GHz, or downlink of X-band satellite communication (7.25-7.75 GHz). Therefore, UWB antennas with band-notched characteristics to filter the potential interference are desirable. Nowadays, to mitigate this effect, many UWB antennas with various band-notched properties have developed [8-11]. Many techniques are also used to introduce notch band for rejecting the interference in the UWB antennas. It is done either by inserting switchable structures in the ground plane [12], applying a shunt open-circuited stub [13], using Step-Impedance Resonators (SIR) [14], inserting fractal structures [15], or using T-ring slot at the square radiating patch [16].

All of the above methods are used for rejecting a single band of frequencies. However, to effectively utilize the UWB spectrum and to

improve the performance of the UWB system, it is desirable to design the antenna with dual band rejection. It will help to minimize the interference between the narrow band systems with the UWB system. Some methods are used to obtain the dual band rejection in the literature [17-20].

In this paper, a different method is proposed to obtain the dual band rejection of frequency bands for HiperLAN and downlink frequency of X-band satellite communication systems. The HiperLAN frequency band is rejected by inserting a G-shaped SRR in the feed-line at center frequency of 5.2 GHz, and a G-shaped SIR is inserted to reject the desired downlink band of satellite systems at a center frequency of 7.5 GHz. The proposed antenna with the dual band-notched function is successfully implemented and the simulation results show reasonable agreement with the measurement results. The designed antenna has a small size. Good VSWR and radiation pattern characteristics are obtained in the frequency band of interest.

## II. ANTENNA DESIGN

The structure of proposed monopole antenna fed by a microstrip line is shown in Fig. 1. The dielectric substance (FR4) with thickness of 1.6 mm with relative permittivity of 4.4 and loss tangent 0.018 is chosen as substrate to facilitate printed circuit board integration. The basic monopole antenna structure consists of a circular radiating patch, a feed line, and a ground plane. The proposed antenna is connected to a 50-Ω SMA connector for signal transmission. The radiating patch is connected to a feed line of width  $W_f$  and length  $L_{gnd}$ . The width of the microstrip feed line is fixed at 2 mm, as shown in Fig. 1. On the other side of the substrate, a conducting ground plane of width  $W_{sub}$  and  $L_{gnd}$  length is placed. Final parameter values of the presented antenna design are specified in Table 1.

Table 1: Final parameter values of the antenna

Parameter	$W_{sub}$	$W_{sub}$	$W$	$L$	$W_f$
(mm)	12	18	0.25	3.8	2
Parameter	$L_f$	$W_1$	$L_1$	$W_2$	$L_2$
(mm)	7	0.25	3.3	0.75	4.25
Parameter	$W_3$	$L_3$	$W_4$	$L_4$	$W_5$
(mm)	0.25	4.75	1.3	0.25	0.75
Parameter	$L_5$	$W_6$	$L_6$	$d$	$R$
(mm)	0.25	1.25	1.65	0.25	5

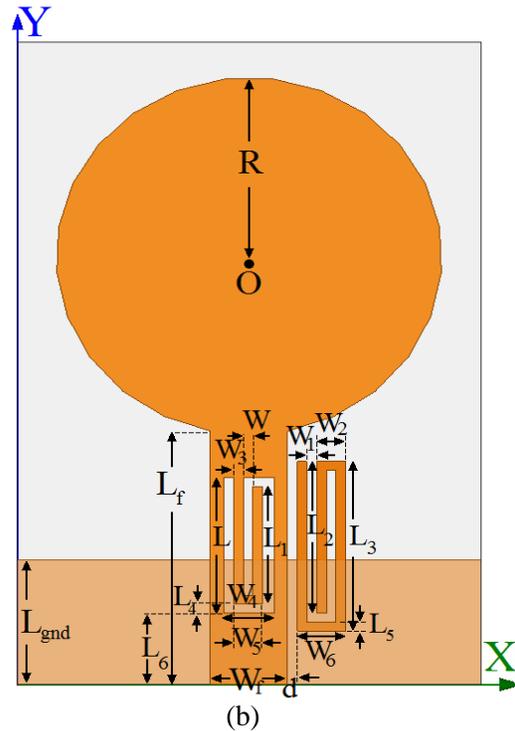
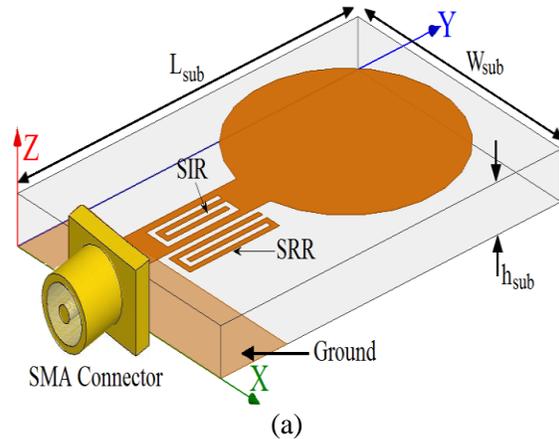


Fig. 1. Structure of the proposed antenna: (a) side view, and (b) top view.

At the first notched frequency, the current concentrated on the edges of the interior and exterior of G-shaped SRR. Additionally, the G-shaped SIR acts as a filtering element to generate another notched frequency, because it can create additional surface current path around of feed-line. At this notched frequency (7.5 GHz), the current flows are more dominant around the modified SIR, and they are oppositely directed between the inserted slits and the radiating patch [8-12]. As a result, the desired high attenuation near the notched frequencies can be produced.

### III. RESULTS AND DISCUSSIONS

In this section, the microstrip monopole antenna with various design parameters was constructed, and the numerical and experimental results of the input impedance and radiation characteristics are presented and discussed. The analysis and performance of the proposed antenna is explored by using Ansoft simulation software High-Frequency Structure Simulator (HFSS) [21], for better impedance matching.

The structure of the various antennas used for simulation studies were shown in Fig. 2. VSWR characteristics for the ordinary circular monopole antenna (Fig. 2 (a)), antenna with a G-shaped SRR at feed-line (Fig. 2 (b)), and the proposed antenna structure (Fig. 2 (c)) are compared in Fig. 3.

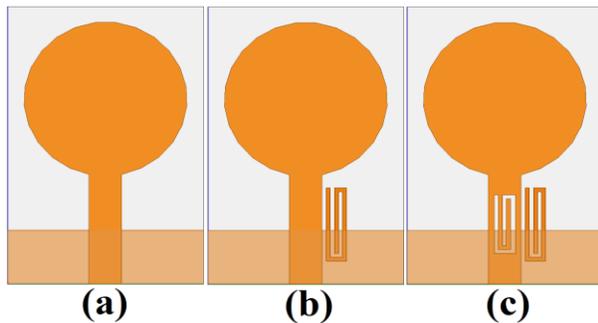


Fig. 2. (a) Ordinary square antenna, (b) antenna with a G-shaped SRR at feed line, and (c) the proposed antenna.

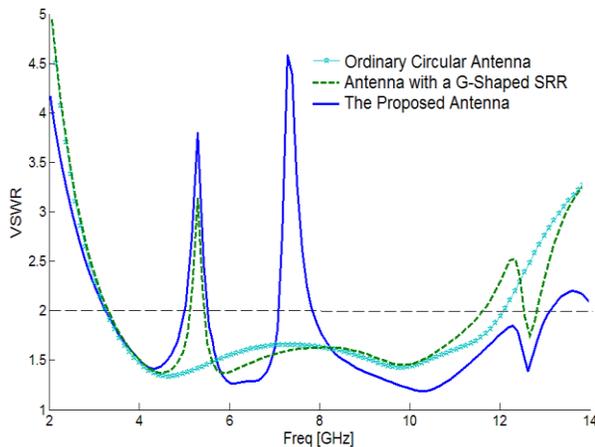


Fig. 3. Simulated VSWR for the various antenna structures shown in Fig. 2.

As shown in Fig. 3, to generate a single frequency band-notched function (5.15-5.35 GHz),

we use a G-shaped SRR, and also by adding a G-shaped SIR at feed-line, the dual band-notched function can be achieved that covers all the 5.2 GHz HiperLAN and 7.5 GHz satellite down-link bands.

The simulated current distribution on the radiating patch for the proposed antenna at the notched frequencies of 5.2 and 7.5 GHz is presented in Figs. 4 (a) and 4 (b), respectively. The current direction on the reject structures is opposite to that on the nearby antenna structure, so the far fields produced by the currents on the reject structures and nearby antenna structure cancel each other in the reject band. In Fig. 4 (b), the current mainly flows around the outer inverted G-shaped SRR, which destructs the radiation of the original antenna at this band and results in the lower stop-band [22-24].

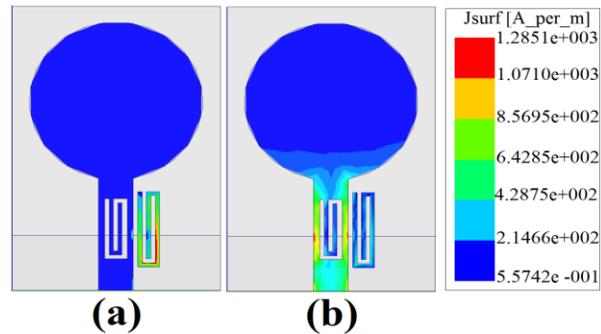


Fig. 4. Simulated surface current distributions for the proposed antenna at the notched frequencies: (a) 5.2 GHz, and (b) 7.5 GHz.

In Fig. 4 (b), the current distribution at 7.5 GHz is even on the outer and inner of G-shaped SIR, but the current direction on the inner slits is opposite to that on the outer ones. It is suggested that two band notches should be mutually interfered with and the notched property does not have any influence on the antenna radiation at 7.5 GHz. Then, a pass-band between two adjacent stop-bands can be obtained because of the mutual interference of the two notched structures [25-27].

Figure 5 shows the measured and simulated VSWR characteristics of the proposed antenna. The fabricated antenna has the frequency band of 3.02 to over 13.32 GHz, with two notch bands around 5.15-5.35 GHz and 7.25-7.75 GHz.

The maximum gain of the proposed antenna was shown in Fig. 6. A sharp decrease of

maximum gain in the notched frequency bands at 5.2 and 7.5 GHz are shown in Fig. 6. For other frequencies outside the notched frequency band, the antenna gain with the filters is similar to those without them [28-32].

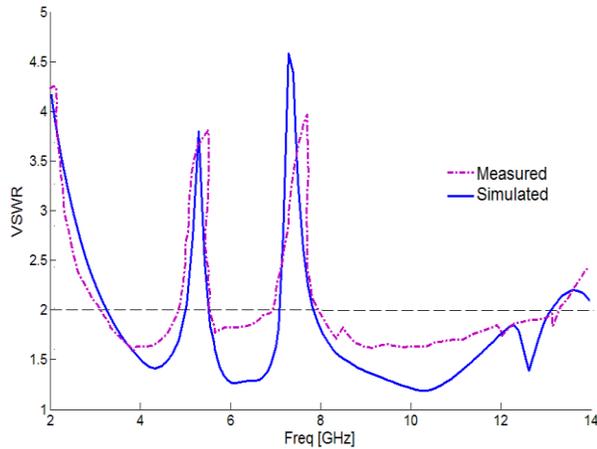


Fig. 5. VSWR comparison of the proposed antenna.

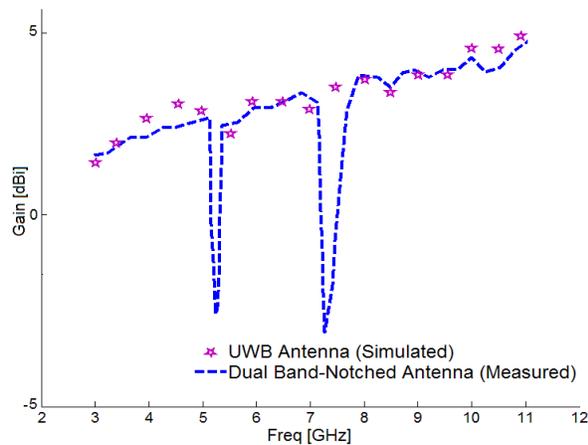


Fig. 6. Measured maximum gain of the proposed antenna.

Figure 7 depicts the measured and simulated radiation patterns including the co-polarization and cross-polarization in the H-plane (x-z plane) and E-plane (y-z plane). It can be seen that nearly omni-directional radiation pattern can be observed on x-z plane over the whole UWB frequency range, especially at the low frequencies. The radiation patterns on the y-z plane are like a small electric dipole leading to bidirectional patterns in a very wide frequency band. With the increase of

frequency, the radiation patterns become worse because of the increasing effects of the cross-polarization [33-38].

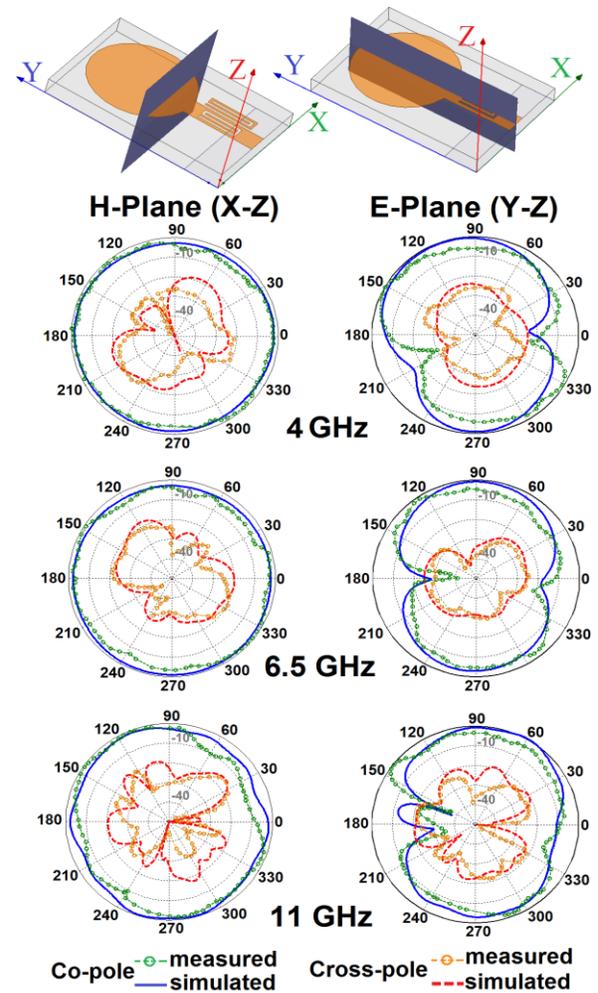


Fig. 7. Measured and simulated radiation patterns.

#### IV. CONCLUSION

In this paper, a new antenna structure is proposed that provides a double stop-band notch in the 5.2/7.5 GHz for various UWB applications. The fabricated antenna has the frequency band of 2.86 to over 12.92 GHz, with two rejection bands around 5.15-5.35 and 7.25-7.75 GHz. The proposed antenna has a simple configuration and small size. The designed antenna can be used in UWB systems to reduce interference between UWB and other wireless communication systems. The proposed antenna has an ordinary square radiating patch, therefore displays a good omni-directional radiation pattern even at higher

frequencies.

### ACKNOWLEDGMENT

The authors are thankful to MWT Company staff for their help (www.microwave-technology.com).

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